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Realising traceable electrostatic forces despite non-linear balance motion

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Abstract. Direct realisation of force, traceable to fundamental constants via electromagnetic balances, is a key goal of the proposed redefinition of the International System of Units (SI). This will allow small force metrology to be performed using an electrostatic force balance (EFB) rather than subdivision of larger forces. Such a balance uses the electrostatic force across a capacitor to balance an external force. In this paper we model the capacitance of a concentric cylinder EFB design as a function of the displacement of its free electrode, accounting for the arcuate motion produced by parallelogram linkages commonly used in EFB mechanisms. From this model we suggest new fitting procedures to reduce uncertainties arising from non-linear motion as well as methods to identify misalignment of the mechanism. Experimental studies on both a test capacitor and the NIST EFB validate the model.

Keywords: Metrology, Electrostatics, Precision measurement, Mass, Force

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1. Introduction

Under the proposed redefinition of the International System of Units (SI), the kilogram will be defined in terms of three fundamental constants (the Planck constant, the speed of light, and the frequency of the radiation corresponding to the transition between the two hyperfine levels of the Caesium-133 atom in its ground state). This will allow measurement of SI traceable mass or force in the laboratory without the use of transfer standards used to link to the International Kilogram Prototype.

This link to fundamental constants has the potential to revolutionise small mass and force metrology. Under the current system, multiple subdivisions from the kilogram must be made to produce small mass transfer standards. Each of these subdivisions adds further compounding uncertainties, progressively limiting the attainable accuracy of mass and force measurements as they decrease in magnitude, with potential adverse consequences

for nanometer-scale science and technology. Once the kilogram is redefined, SI traceable force can be directly realised in the laboratory at any scale, measurements will again be limited by the accuracy of the experiment rather than our knowledge of the unit itself.

The standard method for realising mass directly is the Watt balance. This uses the Lorentz force on a coil of wire in a magnetic field to balance the gravitational force of a test mass by adjusting the current through the wire. Electrostatic forces can also be used to realise force directly from the proposed redefined SI. In practice, large electrostatic forces are more difficult to generate than electromagnetic forces. Thus, at the scale of the kilogram an electrostatic force balance (EFB) is not practical. At small forces a higher relative uncertainty is acceptable, and thus equipment requires less complicated metrology. As such, accurate electrostatic force balances have the potential to realise SI traceable masses and forces at previously unattainable scales. This principle has been used for both small mass metrology [1] and for measuring Newton’s gravitational constant [2].

The basic principle of an EFB is to balance the force under study with a known electrostatic force. The electrostatic force is generated between two electrodes of a capacitor of capacitance C . The force depends on the gradient of the capacitance

$$\mathbf{F} = \frac{1}{2}V^2\nabla C, \quad (1)$$

where V is the voltage across the capacitor. A balance mechanism can be designed so the motion of the two plates is constrained to one dimension, z . Then Equation 1 simplifies to

$$F_z = \frac{1}{2}V^2\frac{\partial C}{\partial z}. \quad (2)$$

By choosing an appropriate capacitor geometry the capacitance gradient in the z direction can be linear. For example a concentric cylinder capacitor where the cylinder axes are aligned with the z -axis.

The sources of uncertainty in such a balance are the accuracy with which the voltage and the capacitance gradient can be measured, along with any further uncertainties introduced by the balance mechanism and its kinematics. To realise mass rather than force the local gravity must also be measured. This can be done, however, using an absolute gravimeter with relative uncertainties of order 10^{-8} .

Attainable uncertainties in voltage, capacitance, and distance will not be the limiting factors in small force metrology. Voltage realisations can achieve relative uncertainties of order 10^{-11} [3], by direct link to quantum effects. Using the calculable capacitor, capacitance can be realised with relative uncertainties of 15×10^{-9} [4]. Interferometric distance measurement achieve absolute uncertainties measured in attometers [5] for atomic scale measurements, and relative uncertainties 4×10^{-10} for a range of a few centimeters [6]. In practice, a balance designed to operate at the 1 mg ($\approx 10 \mu\text{N}$) level—approximately 6 orders of magnitude lower force than the Watt balance—requires a capacitor where $\frac{\partial C}{\partial z}$ is 1 pF/mm to operate at 100 V. At these scales commercially available equipment [7]—used

in the National Institute of Standards and Technology (NIST) EFB—can achieve relative uncertainties of 5×10^{-6} , 15×10^{-8} , and 4×10^{-6} for capacitance [8], distance [9], and voltage, respectively [10]. Thus, if further uncertainties could be avoided this would allow realisation of the milligram with an uncertainty of order 1×10^{-5} . By further calibration of the multimeter and capacitance bridge, combined with careful experimental design these uncertainties could be reduced an order of magnitude further.

Current calibrations for the milligram, relying on subdivision from a kilogram artifact, have relative uncertainties of order 2×10^{-4} [11]. Thus, an electrostatic force balance could reduce uncertainties at this level by two orders of magnitude, provided no additional uncertainties accrue. Both Type A uncertainties from sources such as the stability of the control loop used to control the balance position, and Type B uncertainties from issues such as alignment need to be considered in such a measurement. In this paper we concentrate on Type B uncertainties resulting from the alignment of the electrodes and the electrode translation mechanism, and how this relates to the ability to measure the capacitance gradient. For this we analyse a concentric cylinder capacitor such as used in the NIST electrostatic force balance [1].

2. Theory

As previously mentioned, a cylindrical capacitor design can be used to generate electrostatic forces. To remove any off-axis forces the electrodes should be concentric. For a truly concentric cylinder capacitor with electrodes overlapping by H in the z (axial) direction, the capacitance per unit length can be calculated as

$$\frac{C}{H} = \frac{\epsilon_r}{2k_e \log\left(\frac{a_2}{a_1}\right)}. \quad (3)$$

where a_1 and a_2 are the radii of the inner and outer cylinder respectively. k_e is the Coulomb constant ($= \frac{1}{4\pi\epsilon_0}$), and ϵ_r is the relative permittivity of the medium between the electrodes.

For an accurate force measurement two different readings must be taken. Firstly, the capacitance gradient should be measured by recording the capacitance as the balance position is varied. Secondly a differential recording of the voltage measurement is performed, to measure the difference in the voltage needed to maintain the balance's null point with and without an applied force. These two results can be combined to a final force measurement using Equation 2.

This method assumes the free electrode remains axially aligned and concentric with the fixed electrode throughout its entire range of motion, resulting in a linear capacitance gradient. If this is not the case higher-order terms will be present in $C(z)$, the capacitance with respect to position. If force measurements are performed differentially at $z = 0$ then they are insensitive to any higher order terms. The capacitance gradient of the balance at its null position ($C'(0)$), however, is determined from capacitance measurements at a range

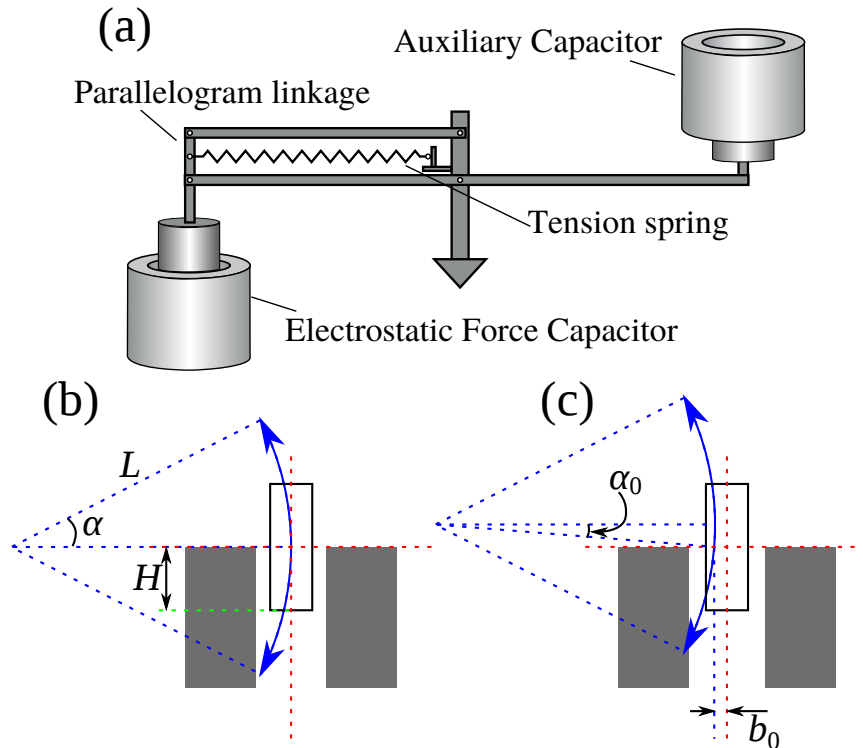


Figure 1. a) Schematic of the NIST electrostatic force balance. The capacitor used to generate traceable electrostatic forces is attached to a parallelogram linkage. The schematic also shows the tension spring used to control the balance stiffness and the auxiliary capacitor used to control the balance motion when measuring the capacitance gradient of the main capacitor. b) Diagram showing the arcuate motion of the inner cylinder in ideal alignment. L is the length of the linkage arms, α is the angle of the linkage, and H is the z -direction overlap of the electrodes when $z = 0$. c) Diagram showing possible misalignment of the capacitor cylinders. b_0 and α_0 are the distance between the electrode axes and the angle of the balance arm at $z = 0$ respectively.

of z positions. Any higher terms in the capacitance vs. z -position must be accounted for in when fitting the data to correctly determine the capacitance gradient at $z = 0$. As such it is essential to know the functional form of $C(z)$.

In practice, a balance mechanism which constrains the electrodes to be both axially aligned and concentric throughout its motion is a considerable challenge. A standard pan balance design uses a balance beam with the balance pans freely-suspended; gravity maintains the balance pan's angle as the balance beam tilts. For an EFB free suspension is not possible; while the off axial forces on a truly concentric capacitor are zero, a freely suspended electrode would be in an unstable equilibrium. This problem is solved by rigidly attaching the electrode to a parallelogram linkage. The parallelogram linkage keeps the axial alignment of the cylinders constant throughout its motion. Both this linkage and a balance beam cannot keep the electrodes concentric, instead there would be lateral displacement proportional to the cosine of the balance tilt. For other types precision balances, such as

a Watt balance, lateral displacement is avoided by suspending balance pans from a band attached to a balance wheel. For an EFB such a mechanism would again introduce an unstable equilibrium.

Both the NIST EFB [1] and other EFBs [12] implement the parallelogram linkage as the cylinder can be aligned to be concentric at the null position maintained during force measurements. In this paper we consider the functional form of $C(z)$ arising from the resulting arcuate motion of the inner electrode in an aligned state, and the effects of misalignment (Figures 1a and b respectively). An aligned state is defined at $z = 0$ when the balance arms are horizontal, and the electrodes are concentric.

Considering first the effect of eccentricity. If the central axes of the two electrodes are separated by a distance b the capacitance per unit length becomes [13]

$$\frac{C}{H} = \frac{\epsilon_r}{2k_e \log \left(\frac{\sqrt{((a_2 - a_1)^2 - b^2)((a_2 + a_1)^2 - b^2)} - b^2 + a_2^2 + a_1^2}{2a_1 a_2} \right)}, \quad (4)$$

This form can be shown to be consistent with that given in Smythe [14] by considering that $\operatorname{arcosh}(x) = \log(x + \sqrt{x^2 - 1})$, and in the case of $b = 0$ reduces to Equation 3.

Taking the Taylor expansion of Equation 4 about $b = 0$ gives

$$\frac{C}{H} = \frac{\epsilon_r}{2k_e} \left(\frac{1}{\log \left(\frac{a_2}{a_1} \right)} + \frac{b^2}{\xi} + O(b^4) \right), \quad (5)$$

where $\xi = (a_2 - a_1)(a_2 + a_1) \log^2 \left(\frac{a_2}{a_1} \right)$. This series converges quickly for $a_1 \gtrsim (a_2 - a_1)$ and $b^2 \ll \frac{4(a_2 - a_1)^2}{3}$.

Defining H as the z -direction overlap at $z = 0$, in the aligned case, Equation 4 simply expands to

$$C(z) = C_0 + \frac{\epsilon_r}{2k_e} \left(\frac{z}{\log \left(\frac{a_2}{a_1} \right)} - \frac{Hz^4 + z^5}{4L^2\xi} + O(z^8) \right), \quad (6)$$

by also using the first two terms is the expansion for b in terms of z . Here C_0 is the capacitance at $z = 0$. We note that despite the fifth order functional form no second or third order terms are present. However, introducing the misalignments b_0 and α_0 from Figure 1b results in a significantly more convoluted functional form:

$$C(z) = C_0 + \frac{\epsilon_r}{2k_e} \left(\frac{z}{\log \left(\frac{a_2}{a_1} \right)} - \frac{(b_0^2 + 2Hb_0 \tan(\alpha_0))z + (2b_0 \tan(\alpha_0) + H \tan^2(\alpha_0))z^2 + \tan^2(\alpha_0)z^3}{\xi} \right. \\ \left. + \frac{Hb_0z^2 - (H \tan(\alpha_0) - b_0)z^3 - \tan(\alpha_0)z^4}{L \cos(\alpha_0)\xi} - \frac{Hz^4 + z^5}{4L^2 \cos^2(\alpha_0)\xi} + O(z^6) \right). \quad (7)$$

Both C_0 and the first order term depend on b_0 and α_0 , and these misalignments also introduce second and third order terms.

3. Experimental

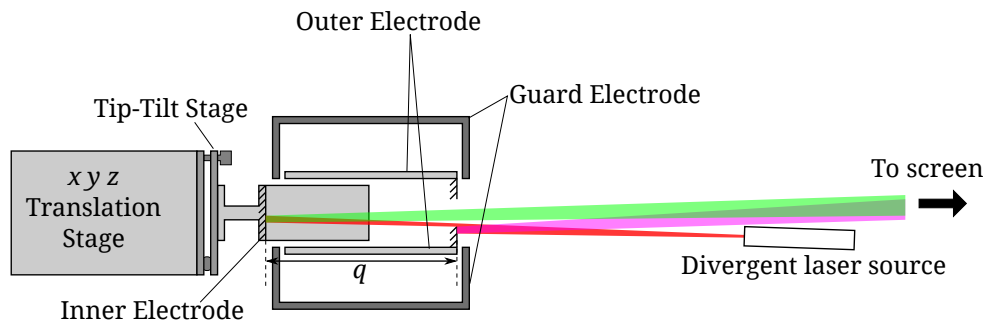


Figure 2. Schematic of the experimental test setup. Three axis encoded translation stages move the inner electrode with respect to the outer electrode of the test capacitor. The capacitor was centred axially by minimising the capacitance between the electrodes. A simple optical lever and a divergent laser source were used for angular alignment of the electrode axes, which was manually adjusted using a tip-tilt stage.

The appropriate order of a polynomial fit to a capacitance gradient could be determined empirically by observing the residuals arising from lower order fits. It is, however, disconcerting to apply high order polynomial fits to data without good theoretical grounding for the higher terms. With enough free parameters one can fit a curve to any data set. One could also use orthogonal polynomials to produce a more robust fit at lower order. The disadvantage of this for the EFB is the gradient of the higher order terms is no-longer zero at $z = 0$. As such all terms must be combined to determine the gradient at $z = 0$, rather than the first order term and its associated uncertainty from covariance.

To experimentally validate the necessity for higher order fits when determining $C'(0)$ we have constructed a capacitor from a pair of cylindrical electrodes. The inner electrode was held at ground and the outer electrode was fitted with a grounding guard electrode on top to limit stray capacitance. The diameters of the inner and outer electrodes were nominally 22 mm, and 23 mm respectively. The cylindrical electrodes were manufactured with a reflective surface parallel to the central axis to a machine tolerance of $10 \mu\text{rad}$. The inner electrode was attached to a Newport 562-XYZ 3-axis stage, controlled with three Newport VP-25AA high-precision motorised actuators with 100 nm resolution, to simulate balance motion. An Andeen-Hagerling 2700A bridge measured capacitance between the electrodes at a frequency of 1 kHz and a maximum potential of 15 V. All measurements were performed at a temperature of $20 \pm 0.5^\circ\text{C}$, a pressure of 1 atmosphere, and a relative humidity of 40 %.

The angle between the electrodes was measured using a simple optical lever technique (Figure 2). A divergent laser beam was split between mirrors on the inner and outer electrode, the reflected beams were projected on to a screen 2.69 m from the mirror of the outer electrode. At the screen the laser spot sizes were ≈ 7 mm in diameter. As the distance from the inner electrode mirror to the outer electrode mirror (q) was varied between 80 mm

Table 1. Experimental parameters with Type A uncertainties ($k = 1$)

Parameter	Value	Source
C_0	19.846116(61) pF	Calculated fit
C'	1.116902(32) nF/m	
a_1	11.023(4) mm	Measured with encoded translation stage.
$\frac{a_1}{a_2}$	1.051105(30)	Using Eq. 3.
a_2	11.586(4) mm	From a_1 and $\frac{a_1}{a_2}$.

and 64 mm, no visible changes to the spot positions were recorded. The angle between the incident and reflected beams was 3° , thus this should separate the final spots by ≈ 4 mm. For the optical lever length used, a systematic spot separation of 4 mm results in a 1.5 mrad misalignment. Taking into account spot size, an uncertainty of 3 mrad was associated with angular alignment.

Before each sweep in z , the inner electrode was centred in x and y at $z = -8$ mm and

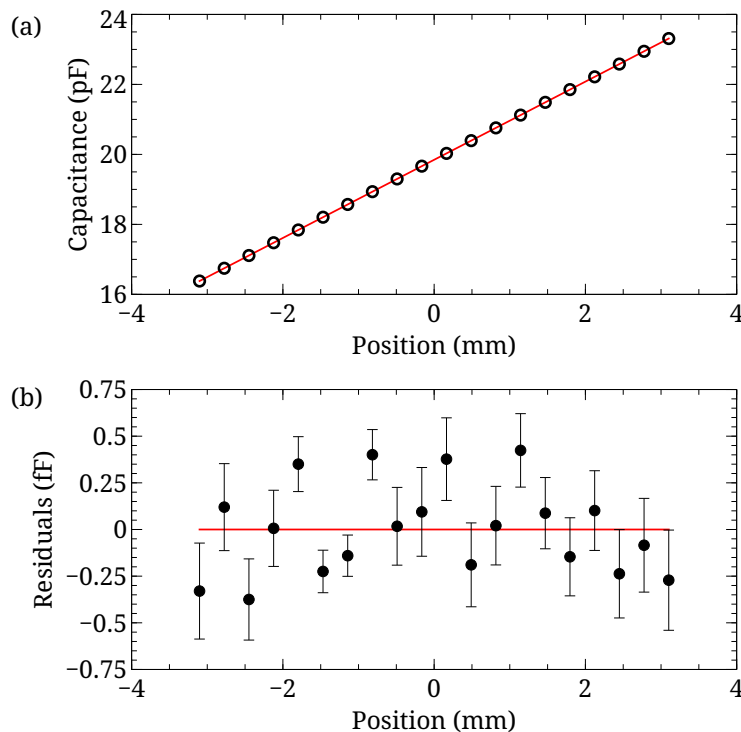


Figure 3. a) Capacitance gradient for linear motion of the inner electrode. These results were used to verify the alignment of the test setup and to determine the capacitance gradient. b) Residuals from a first order fit are presented. These show some minimal sign of a parabolic structure which indicate slight misalignment of the cylinder axes. No residual are greater than 500 aF. For comparison a 3 mrad misalignment of the motion axis would result in a 520 aF deviation from linearity at $z = 3.2$ mm. Error bars represent one standard deviation from repeated measurements.

8 mm by minimising the capacitance, as described previously. The central z axis was then defined between these two points; from this axis the motion to sweep in x , y , and z was calculated. As large motions in z amplify unintended misalignments, higher order terms, and capacitor edge effects, sweep data from $z = \pm 3.2$ mm was analysed. This 6.4 mm range of motion is still large compared to the range of motion of the NIST EFB which is approximately 2.2 mm.

As an initial test of the experimental set-up and alignment, a capacitance gradient was taken using linear motion with the central axes of the two electrodes aligned. The data, shown in Figure 3, was fitted with a first order polynomial, $C = C_0 + C'z$. The fit's adjusted coefficient of determination differs from unity by a few parts per billion ($1 - \bar{R}^2 = 1.4 \times 10^{-8}$). All data points lie within 500 attofarad of the fit. The fit results and derived parameters are shown in Table 1.

First considering the aligned case where at $z = 0$ the electrodes are concentric and the arc angle is zero (Figure 1a); no second or third order polynomial terms should be present in $C(z)$. Without changing the alignment from the linear measurements the inner electrode was set to move over an arc with a maximum deviation of 0.2 mm at ± 8 mm. This is consistent

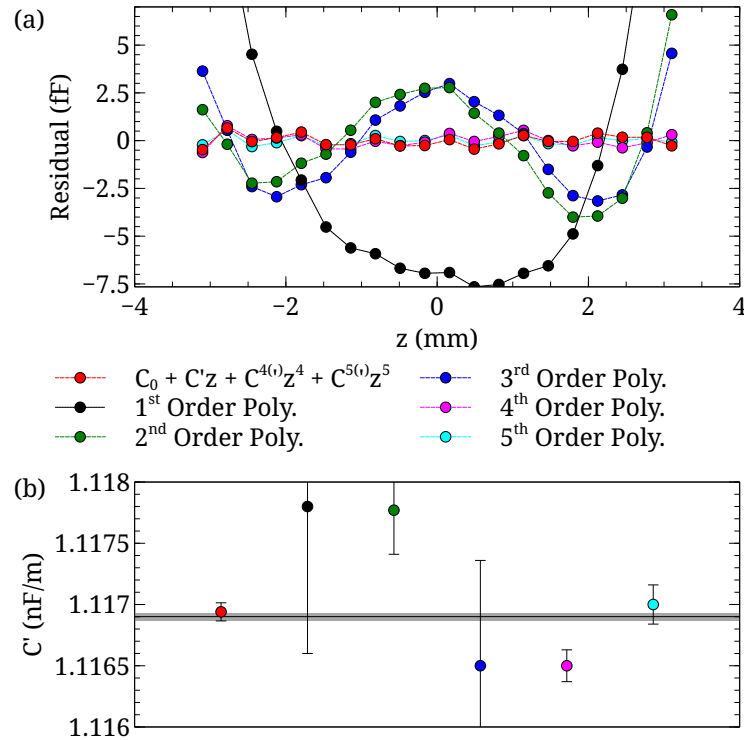


Figure 4. a) Fit residuals for a $C(z)$ measured with arcuate balance motion. Fits performed for different polynomial models. b) First order capacitance gradient, $C'(0)$, extracted from each fit. Error bars represent $k = 1$ type A uncertainty estimated from least squares fitting. The horizontal line and grey area represent the capacitance gradient and type A uncertainty from the linear motion presented in Figure 3.

with a beam length of $L = 160.1$ mm for a 4-bar linkage.

One method to validate the model would be to use the parameters in Table 1 to check the magnitude of higher order terms. However, the fit parameters for the higher order terms are very sensitive to any measurement uncertainties, and the accuracy to which they can be measured is of little interest for electrostatic forces generated at $z = 0$. Thus, we instead fit our model and the first five orders of polynomial fits to the data and compare the returned value for the gradient at $z = 0$ to the gradient measured for the linear motion (Figure 4).

From Figure 4 it is clear that the model of fourth and fifth order terms added to the linear capacitance gradient (henceforth, referred to as a reduced-fifth-order polynomial) returns the same result as linear motion. The odd-order polynomial fits also return the correct result within the uncertainty from the linear least squares fit, but with significantly larger uncertainty than the reduced-fifth-order polynomial model. The second and fourth order polynomial fits, however, return values more than two sigma from the linear result. It is worth noting that the residuals of the fourth order polynomial are not significantly larger than that of the reduced-fifth-order polynomial nor the fifth order polynomial, nor do they display a clear structure.

Our model shows that the introduction of an offset in the $x - y$ plane, such that the two electrodes are not concentric at $z = 0$, introduces second and third order terms in z to the capacitance sweep. Also the capacitance gradient at $z = 0$ will increase as the electrodes are no longer concentric, the condition for a capacitance gradient minimum. For this test a smaller amplitude of the arcuate motion was used, in this case corresponding to a parallelogram linkage with length $L = 320.05$ mm. The centre of the motion was laterally offset by $\pm 50 \mu\text{m}$. As expected, Figure 5 shows that the reduced-fifth-order polynomial model is no longer a good fit to the data once the offset is introduced. As such a full fifth-order polynomial is necessary to account for possible misalignments. Inspecting the reduced-fifth-order fit residuals could potentially lead to a method for checking misalignment. However, misalignments from other sources, such as $\alpha_0 \neq 0$ will also cause similar residuals, thus determining the exact source of misalignment may be difficult in practice.

4. Discussion

The presented results agree with the theoretical model, and confirm that the arcuate motion induced by parallelogram linkages on EFBs can be accurately accounted for if the correct model is used to when fitting $C(z)$. Comparing uncertainties in $C'(0)$ for our test setup for linear and arcuate motion shows that the relative uncertainty increases from 29×10^{-6} in the linear case to 66×10^{-6} and 143×10^{-6} for the reduced-fifth-order polynomial and fifth-order polynomial fits respectively for arcuate motion. The increased uncertainty arises from the extra degrees of freedom of the fit.

It is important to note that the magnitude of the relative $C'(0)$ uncertainties presented in this paper are not of particular significance. The experiment was designed to easily compare linear and non-linear motion, rather than to produce the lowest uncertainty. As such it

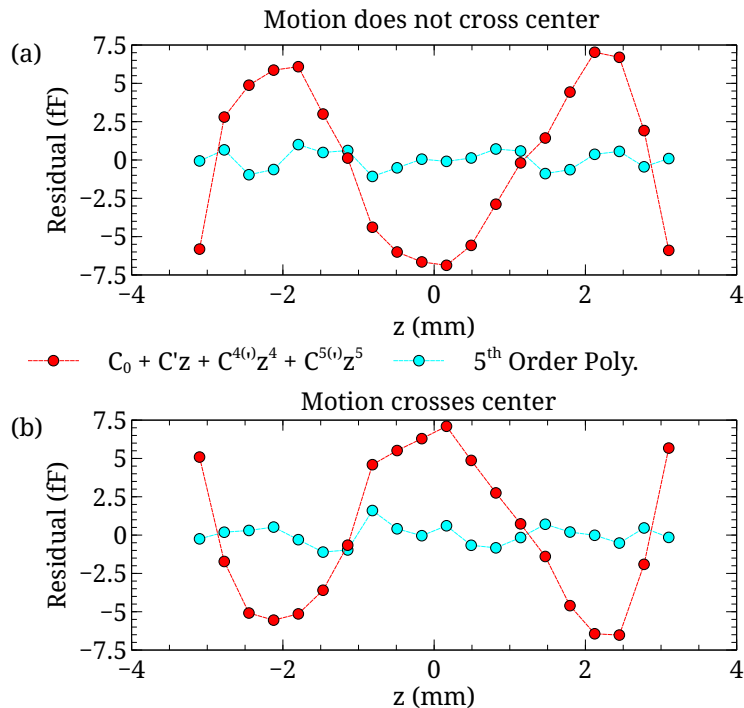


Figure 5. Comparison of fifth-order and reduced fifth-order polynomial fit residuals for $C'(z)$ when the arcuate motion is displaced $50 \mu\text{m}$ in the $x - y$ plane. The displacement is parallel to the parasitic $x - y$ motion of the simulated mechanism. The displacement was repeated in each direction, both towards the simulated balance mechanism (as such the arcuate motion will never centrally align the cylinders) and away from the simulated mechanism. As expected, the reduced fifth-order polynomial performs poorly under misalignment.

is worth considering results from the NIST EFB. In Figure 6a we compare the calculated values for $C'(0)$ from 6 capacitance gradient determinations taken over 12 days in vacuum on the EFB. Capacitance was measured at 11 discrete z positions over a nominal range of ± 1 mm. To avoid polarisation mixing errors in the double-pass Michelson interferometer used to measure displacement, each z position used is a multiple of one quarter of the wavelength of the laser. Capacitance readings were performed with an Andeen-Hagerling 2500A bridge traceably calibrated to the NIST calculable capacitor. For more details on the instrumentation of the NIST EFB see [1, 15].

The data in Figure 6a was not fitted with the reduced-fifth-order polynomial as the residuals plotted in Figure 6b make it clear that the second and third order terms are needed to correct for misalignment. These residuals are qualitatively very similar to those presented in Figure 5 except notably the amplitude is almost three orders of magnitude smaller, with a maximum absolute residual of only 21 aF in for the reduced-fifth-order case, and 2 aF for the fifth order fit. This approaches the 0.5 aF resolution of the capacitance bridge.

Figure 6a compares odd order polynomial fitting up to the suggested fifth order fit.

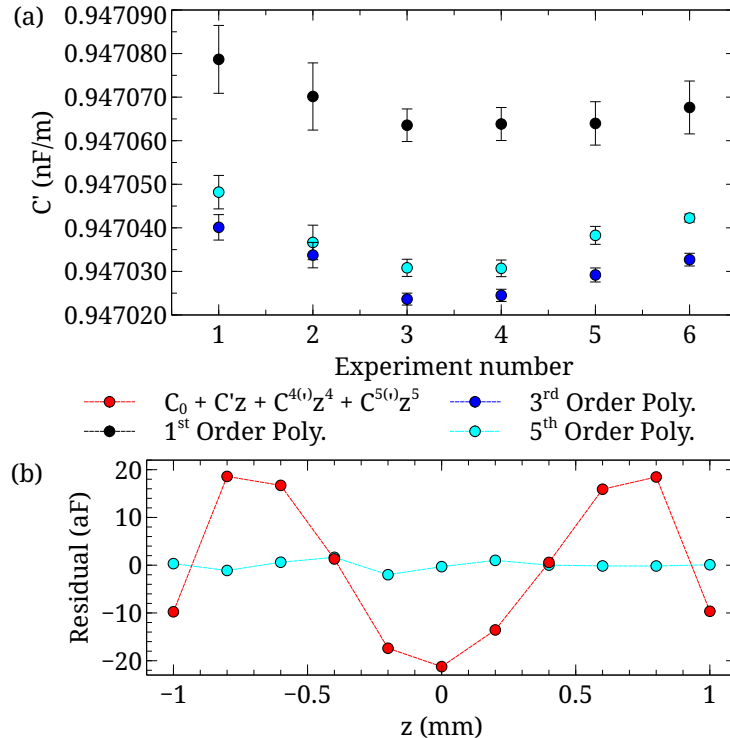


Figure 6. a) Shows a series of capacitance gradient determinations performed on the NIST EFB over 12 days. These results show a systematic difference when lower order fits are used to determine the gradient. Error bars represent $k = 1$ type A uncertainties estimated from least squares fitting. The reduced fifth-order fit was not used as the fit residuals shown in b) reveal that misalignment similar to Figure 5. The magnitude of this misalignment is, however, much smaller than that presented above. The fit residuals from the first and third order polynomial are not presented as they dwarf the effect being shown.

The results reveal a relative systematic decrease in $C'(0)$ of approximately 30×10^{-6} when using fifth order polynomial fitting rather than a simple linear fit. This discrepancy exceeds type A uncertainty from the linear fit at 5σ . Third order fitting compared to 5th results in a relative effect of 8×10^{-6} which, while outside the type A uncertainty for the measurement, is smaller than the measurement drift between results. While a relative uncertainty of 30×10^{-6} may be dominant for some EFB results, this effect is highly dependent on the distance the electrodes are swept, and the electrode alignment, so should not be considered a systematic uncertainty which can retroactively be applied.

5. Conclusion

We have derived a model for concentric cylinder electrostatic force balances using parallelogram linkages and verified it experimentally. The model shows 5th order polynomial terms in the capacitance gradient used to realise a traceable force, and that lower order fits can result in a systematic error in the measured force. The magnitude of this error

depends on a number of factors such as the dimensions of the instrument, the distance moved in the capacitance gradient determination, and the balance alignment. For measurements performed on the NIST EFB the higher order fits resulted in a relative reduction of $C'(0)$ (and hence the measured force) by 30×10^{-6} compared to a linear fit. We note that this should not be considered a fixed systematic for the apparatus as it depends on both the current alignment and the displacement over which the capacitance is measured. It does, however, confirm that as future work pushes micro/nano Newton force metrology to the parts per million level the full functional form of the $C(z)$ must be considered.

6. References

- [1] Jon R Pratt and John A Kramar. SI realization of small forces using an electrostatic force balance. In *Proceedings of the 18th Imeko World Congress*, 2006.
- [2] Terry Quinn, Clive Speake, Harold Parks, and Richard Davis. The BIPM measurements of the Newtonian constant of gravitation, G . *Philosophical Transactions of the Royal Society A: Mathematical, Physical and Engineering Sciences*, 372(2026):20140032–20140032, 2014.
- [3] B M Wood and S Solve. A review of Josephson comparison results. *Metrologia*, 46(6):R13–R20, dec 2009.
- [4] A.M. Jeffery, L.H. Lee, and J.Q. Shields. Model tests to investigate the effects of geometrical imperfections on the NIST calculable capacitor. *IEEE Transactions on Instrumentation and Measurement*, 48(2):356–359, apr 1999.
- [5] E Massa, G Mana, U Kuetgens, and L Ferroglio. Measurement of the $\{2\ 2\ 0\}$ lattice-plane spacing of a 28 Si x-ray interferometer. *Metrologia*, 48(2):S37–S43, apr 2011.
- [6] John R Lawall. Fabry-Perot metrology for displacements up to 50 mm. *Journal of the Optical Society of America. A, Optics, image science, and vision*, 22(12):2786–2798, 2005.
- [7] Certain commercial equipment, instruments, or materials (or suppliers, or software, etc.) are identified in this paper to foster understanding. Such identification does not imply recommendation or endorsement by the National Institute of Standards and Technology, nor does it imply that the materials or equipment identified are necessarily the best available for the purpose.
- [8] Manufacturer specifications for Andeen-Hagerling AH 2500A. From <http://www.andeen-hagerling.com/ah2500a.htm>, accessed 7 December 2015.
- [9] Manufacturer specifications for Zygo ZMI 4104. From <http://www.zygo.com/met/markets/stageposition/zmi/measboards/4104.pdf>, accessed 7 December 2015.
- [10] Manufacturer specifications for Keysight 3458A. From <http://www.keysight.com/en/pd-1000001297%3Aeps%3Apro-pn-3458A/>, accessed 7 December 2015.
- [11] Z. J. Jabbour and S. L. Yaniv. The kilogram and measurements of mass and force. *Journal Of Research Of The National Institute Of Standards And Technology*, 106(1):25–46, 2001.
- [12] In-Mook Choi, Min-Seok Kim, Sam-Yong Woo, and Soo Hyun Kim. Parallelism error analysis and compensation for micro-force measurement. *Measurement Science and Technology*, 15(1):237–243, jan 2004.
- [13] Chester Snow. Formulas for Computing Capacitance and Inductance. *National Bureau of Standards Circular*, 544:1, 1954.
- [14] William Ralph Smythe. *Static and dynamic electricity*. McGraw-Hill New York, 1950.
- [15] Gordon A Shaw, Julian Stirling, John A Kramar, Alexander Moses, Patrick Abbott, Richard Steiner, Andrew Koffman, Jon R Pratt, and Zeina J Kubarych. The electronic milligram. *Metrologia*, 53:A86, 2016.